
Rate Distortion Theory & Quantization

- Rate Distortion Theory
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- Scalar Quantization
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- High Resolution Approximations
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- Vector Quantization

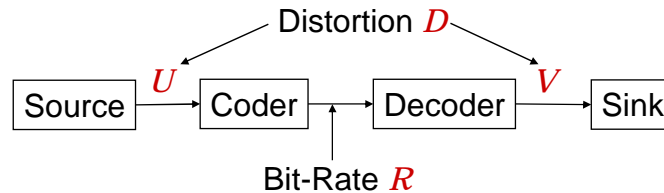


Rate Distortion Theory

- Theoretical discipline treating data compression from the viewpoint of information theory.
- Results of rate distortion theory are obtained without consideration of a specific coding method.
- **Goal:** Rate distortion theory calculates minimum transmission bit-rate R for a given distortion D and source.



Transmission System



- Need to define U , V , Coder/Decoder, Distortion D , and Rate R .
- Need to establish functional relationship between U, V, D , and R



Definitions

- **Source symbols** are given by the random sequence $\{U_k\}$
 - Each U_k assumes values in the discrete set $\mathcal{U} = \{u_0, u_1, \dots, u_{M-1}\}$
 - For a binary source: $\mathcal{U} = \{0, 1\}$
 - For a picture: $\mathcal{U} = \{0, 1, \dots, 255\}$
 - For simplicity, let us assume U_k to be independent and identically distributed (i.i.d.) with distribution $\{P(u), u \in \mathcal{U}\}$
- **Reconstruction symbols** are given by the random sequence $\{V_k\}$ with distribution $\{P(v), v \in \mathcal{V}\}$
 - Each V_k assumes values in the discrete set $\mathcal{V} = \{v_0, v_1, \dots, v_{N-1}\}$
 - The sets \mathcal{U} and \mathcal{V} need not to be the same



Coder / Decoder

- Statistical description of **Coder/Decoder**, i.e. the mapping of the source symbols to the reconstruction symbols, via

$$Q = \{Q(v | u), u \in \mathcal{U}, v \in \mathcal{V}\}$$

- Q is the conditional probability distribution over the letters of the reconstruction alphabet \mathcal{U} given a letter of the source alphabet \mathcal{V}
- Transmission system is described via

Joint pdf : $P(u, v)$

$$P(u) = \sum_{v \in \mathcal{V}} P(u, v)$$

$$P(v) = \sum_{u \in \mathcal{U}} P(u, v)$$

$$P(u, v) = P(u) \cdot Q(v | u) \text{ (Bayes' rule)}$$



Distortion

- To determine distortion, we define a non-negative cost function

$$d(u, v), \quad d(\cdot, \cdot) : \mathcal{U} \times \mathcal{V} \rightarrow [0, \infty)$$

- Examples for d
 - Hamming distance: $d(u, v) = \begin{cases} 0, & \text{for } u = v \\ 1, & \text{for } u \neq v \end{cases}$
 - Squared error: $d(u, v) = |u - v|^2$

- Average **Distortion**

$$D(Q) = \sum_{u \in \mathcal{U}} \sum_{v \in \mathcal{V}} \underbrace{P(u) \cdot Q(v | u)}_{P(u, v)} \cdot d(u, v)$$



Mutual Information

- Shannon average mutual information

$$\begin{aligned} I &= H(U) - H(U|V) \\ &= -\sum_{u \in \mathcal{U}} P(u) \cdot \log P(u) + \sum_{u \in \mathcal{U}} \sum_{v \in \mathcal{V}} P(u, v) \cdot \log P(u|v) \\ &= -\sum_{u \in \mathcal{U}} \sum_{v \in \mathcal{V}} P(u, v) \cdot \log P(u) + \sum_{u \in \mathcal{U}} \sum_{v \in \mathcal{V}} P(u, v) \cdot \log \frac{P(u, v)}{P(v)} \\ &= \sum_{u \in \mathcal{U}} \sum_{v \in \mathcal{V}} P(u, v) \cdot \log \frac{P(u, v)}{P(u) \cdot P(v)} \end{aligned}$$

- Using Bayes' rule

$$I(Q) = \sum_{u \in \mathcal{U}} \sum_{v \in \mathcal{V}} \frac{P(u) \cdot Q(v|u)}{P(u, v)} \cdot \log \frac{Q(v|u)}{P(v)},$$

with $P(v) = \sum_{u \in \mathcal{U}} P(u) \cdot Q(v|u)$



Rate

- Shannon average mutual information expressed via entropy

$$I(U; V) = \underset{\substack{\uparrow \\ \text{Source entropy}}}{H(U)} - \underset{\substack{\uparrow \\ \text{Equivocation: conditional entropy}}}{H(U|V)}$$

- Equivocation:
 - The conditional entropy (uncertainty) about the source U given the reconstruction V
 - A measure for the amount of missing [quantized] information in the received signal V



Rate Distortion Function

- Definition: $R(D^*) = \min_{Q: D(Q) \leq D^*} \{I(Q)\}$
- For a given maximum average distortion D , the rate distortion function $R(D^*)$ is the lower bound for the transmission bit-rate.
- The minimization is conducted for all possible mappings Q that satisfy the average distortion constraint
- $R(D^*)$ is measured in bits for ld



Discussion

- In information theory: maximize mutual information for efficient communication
- In rate distortion theory: minimize mutual information
- In rate distortion theory: source is given, not the channel
- Problem which is addressed:
Determine the minimum rate at which information about the source must be conveyed to the user in order to achieve a prescribed fidelity.
- Another view: Given a prescribed distortion, what is the channel with the minimum capacity to convey the information.
- Alternative definition via interchanging the roles of rate and distortion



Distortion Rate Function

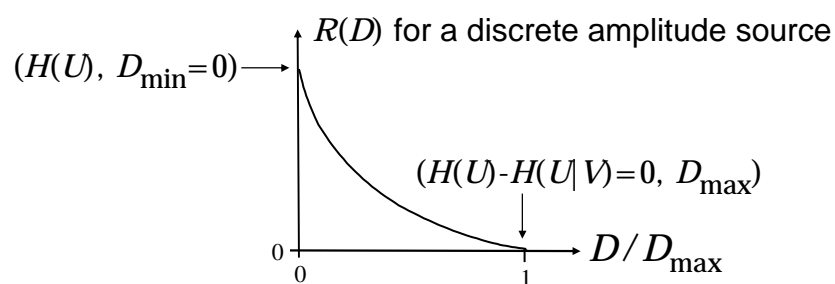
- Definition: $D(R^*) = \min_{Q: I(Q) \leq R^*} \{d(Q)\}$

- For a given maximum average rate R , the distortion rate function $D(R^*)$ is the lower bound for the average distortion.

- Here, we can set R^* to the capacity C of the transmission channel and determine the minimum distortion for this ideal communication system



Properties of the Rate Distortion Function, I



- $R(D)$ is well defined for $D \in (D_{\min}, D_{\max})$
- For discrete amplitude sources, $D_{\min} = 0$
- $R(D) = 0$, if $D > D_{\max}$

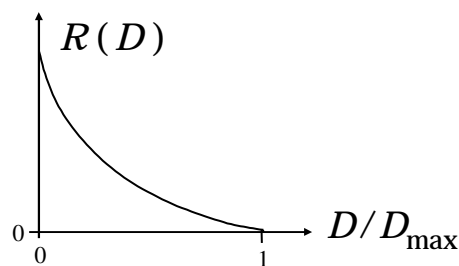


Properties of the Rate Distortion Function, II

- $R(D)$ is always positive

$$0 \leq I(U;V) \leq H(U)$$

- $R(D)$ is non-increasing in D
- $R(D)$ is strictly convex downward in the range (D_{\min}, D_{\max})
- The slope of $R(D)$ is continuous in the range (D_{\min}, D_{\max})



Shannon Lower Bound

- It can be shown that $H(U - V | V) = H(U | V)$

- Then we can write

$$\begin{aligned} R(D^*) &= \min_{Q: D(Q) \leq D^*} \{H(U) - H(U | V)\} \\ &= H(U) - \max_{Q: D(Q) \leq D^*} \{H(U | V)\} \\ &= H(U) - \max_{Q: D(Q) \leq D^*} \{H(U - V | V)\} \end{aligned}$$

- Ideally, the source coder would produce distortions $u-v$ that are statistically independent from the reconstructed signal v (not always possible!).

- Shannon Lower Bound: $R(D^*) \geq H(U) - \max_{Q: D(Q) \leq D^*} H(U - V)$



$R(D^*)$ for a Memoryless Gaussian Source and MSE Distortion

- Gaussian source, variance σ^2
- Mean squared error (MSE) $D = E \{ (u-v)^2 \}$

$$R(D^*) = \frac{1}{2} \log \frac{\sigma^2}{D^*}; \quad D(R^*) = \sigma^2 2^{-2R^*}, R \geq 0$$

$$SNR = 10 \cdot \log_{10} \frac{\sigma^2}{D} = 10 \cdot \log_{10} 2^{2R} \approx 6R \text{ [dB]}$$

- Rule of thumb: 6 dB ~ 1 bit
- The $R(D^*)$ for non-Gaussian sources with the same variance σ^2 is always below this Gaussian $R(D^*)$ curve.



$R(D^*)$ Function for Gaussian Source with Memory I

- Jointly Gaussian source with power spectrum $S_{uu}(\omega)$
- MSE: $D = E \{ (u-v)^2 \}$
- Parametric formulation of the $R(D^*)$ function

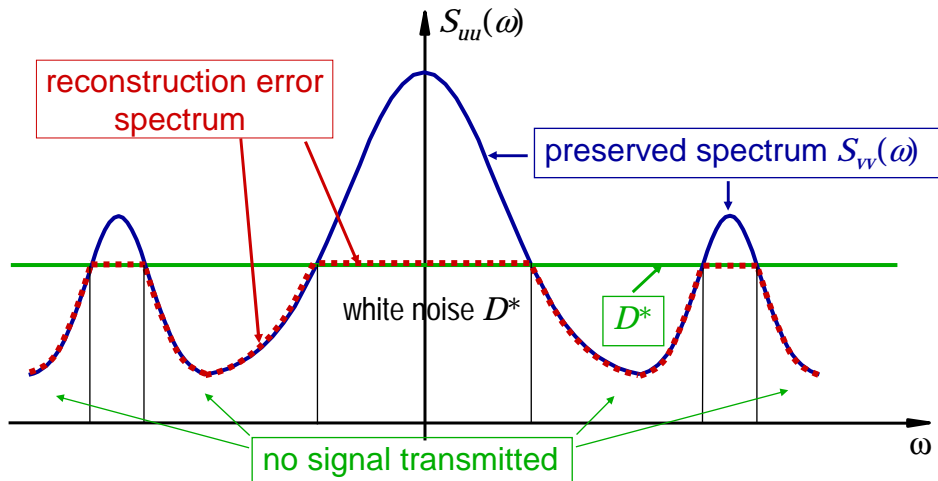
$$D = \frac{1}{2\pi} \int_{\omega} \min[D^*, S_{uu}(\omega)] d\omega$$

$$R = \frac{1}{2\pi} \int_{\omega} \max[0, \frac{1}{2} \log \frac{S_{uu}(\omega)}{D^*}] d\omega$$

- $R(D^*)$ for non-Gaussian sources with the same power spectral density is always lower.



$R(D^*)$ Function for Gaussian Source with Memory II



$R(D^*)$ Function for Gaussian Source with Memory III

- ACF and PSD for a first order AR(1) Gauss-Markov process:

$$U[n] = Z[n] + \rho U[n-1]$$

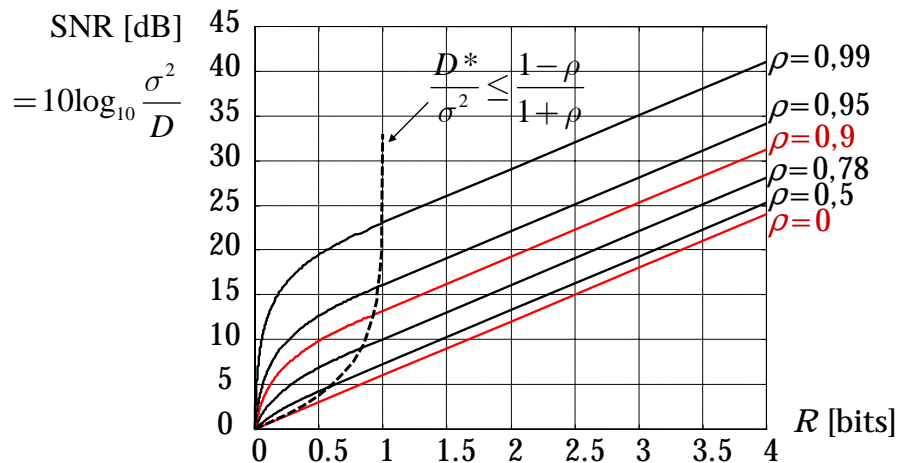
$$R_{uu}(k) = \rho^{|k|} \sigma^2, \quad S_{uu}(\omega) = \frac{\sigma^2(1-\rho^2)}{1-2\rho \cos \omega + \rho^2}$$

- Rate Distortion Function:

$$\begin{aligned} R(D^*) &= \frac{1}{4\pi} \int_{-\pi}^{\pi} \log_2 \frac{S_{uu}(\omega)}{D^*} d\omega, \quad \frac{D^*}{\sigma^2} \leq \frac{1-\rho}{1+\rho} \\ &= \frac{1}{4\pi} \int_{-\pi}^{\pi} \log_2 \frac{\sigma^2(1-\rho^2)}{D^*} d\omega - \frac{1}{4\pi} \int_{-\pi}^{\pi} \log_2 (1-2\rho \cos \omega + \rho^2) d\omega \\ &= \frac{1}{2} \log_2 \frac{\sigma^2(1-\rho^2)}{D^*} = \frac{1}{2} \log_2 \frac{\sigma_z^2}{D^*}. \end{aligned}$$

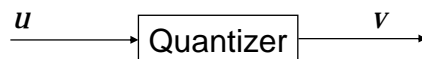


$R(D^*)$ Function for Gaussian Source with Memory IV

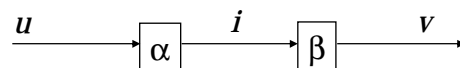


Quantization

- Structure



- Alternative: coder (α) / decoder (β) structure



- Insert entropy coding (γ) and transmission channel



Scalar Quantization

- Average distortion

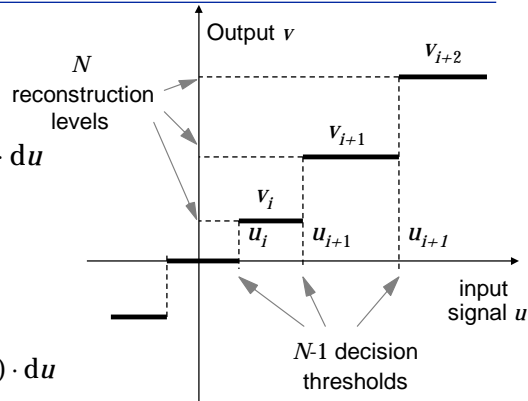
$$D = E\{d(U, V)\}$$

$$= \sum_{k=0}^{N-1} \int_{u_k}^{u_{k+1}} d(u, v_k) \cdot f_U(u) \cdot du$$

- Assume MSE

$$d(u, v_k) = (u - v_k)^2$$

$$D = \sum_{k=0}^{N-1} \int_{u_k}^{u_{k+1}} (u - v_k)^2 \cdot f_U(u) \cdot du$$



- Fixed code word length vs. variable code word length

$$R = \log N \text{ vs. } R = -E\{\log P(v)\}$$



Lloyd-Max Quantizer

- 0: Given: a source distribution $f_U(u)$
a set of reconstruction levels $\{v_k\}$

- 1: Encode given $\{v_k\}$ (Nearest Neighbor Condition):

$$\alpha(u) = \operatorname{argmin} \{d(u, v_k)\} \rightarrow u_k = (v_k + v_{k+1})/2 \text{ (MSE)}$$

- 2: Update set of reconstruction levels given $\alpha(u_k)$
(Centroid Condition):

$$v_k = \operatorname{argmin} E\{d(u, v_k) \mid \alpha(u) = k\} \rightarrow v_k = \frac{\int_{u_k}^{u_{k+1}} u \cdot f_U(u) du}{\int_{u_k}^{u_{k+1}} f_U(u) du} \text{ (MSE)}$$

- 3: Repeat steps 1 and 2 until convergence



High Resolution Approximations

- Pdf of U is roughly constant over individual cells C_k

$$f_U(u) \approx f_k, \quad u \in C_k$$

- The fundamental theorem of calculus

$$P_k = \Pr(u \in C_k) = \int_{u_k}^{u_{k+1}} f_U(u) \cdot du \approx (u_{k+1} - u_k) \cdot f_k = \Delta_k f_k$$

- Approximate average distortion (MSE)

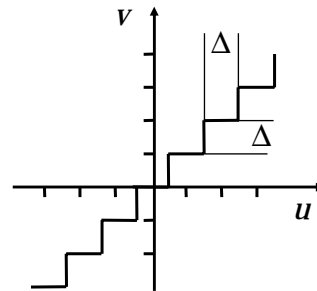
$$\begin{aligned} D &= \sum_{k=0}^{N-1} \int_{u_k}^{u_{k+1}} (u - v_k)^2 \cdot f_U(u) \cdot du = \sum_{k=0}^{N-1} f_k \int_{u_k}^{u_{k+1}} (u - v_k)^2 du \\ &= \sum_{k=0}^{N-1} f_k \frac{\Delta_k^3}{12} = \frac{1}{12} \sum_{k=0}^{N-1} P_k \Delta_k^2 \end{aligned}$$



Uniform Quantization

- Reconstruction levels of quantizer $\{v_k\}$, $k \in K$ are uniformly spaced
- Quantizer step size, i.e. distance between reconstruction levels: Δ
- Average distortion

$$\begin{aligned} \sum_{k=0}^{N-1} P_k &= 1, \quad \Delta_k = \Delta \\ D &= \frac{1}{12} \sum_{k=0}^{N-1} P_k \Delta_k^2 = \frac{\Delta^2}{12} \sum_{k=0}^{N-1} P_k = \frac{\Delta^2}{12} \end{aligned}$$



- Closed-form solutions for pdf-optimized uniform quantizers for Gaussian RV only exist for $N=2$ and $N=3$
- Optimization of Δ is conducted numerically



Panter and Dite Approximation

- Approximate solution for optimized spacing of reconstruction and decision levels
- Assumptions: high resolution and smooth pdf $\Delta(u)$

$$\Delta(u) = \frac{\text{const}}{\sqrt[3]{f_U(u)}}$$

- Optimal pdf of reconstruction levels is not the same as for the input levels
- Average Distortion $D \approx \frac{1}{12N^2} \left(\int_{\mathbb{R}} f_U^{1/3}(u) \cdot du \right)^3$

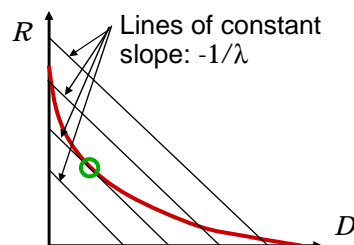
- Operational distortion rate function for Gaussian RV

$$U \sim N(0, \sigma^2), D(R) \approx \frac{\pi\sqrt{3}}{2} \sigma^2 2^{-2R}$$



Entropy-Constrained Quantization

- So far: each reconstruction level is transmitted with fixed code word length
- Encode reconstruction levels with variable code word length
- Constrained design criteria:
 $\min D, \text{ s.t. } R < R_c \text{ or } \min R, \text{ s.t. } D < D_c$
- Pose as unconstrained optimization via Lagrangian formulation:
 $\min D + \lambda R$



- For a given λ , an optimum is obtained corresponding to either R_c or D_c
- If λ small, then D small and R large
- If λ large, then D large and R small
- Optimality also for functions that are neither continuous nor differentiable



Chou, Lookabaugh, and Gray Algorithm*

- 0: Given: a source distribution $f_U(u)$
a set of reconstruction levels $\{v_k\}$
a set of variable length code (VLC) words $\{\gamma_k\}$
with associated length $|\gamma_k|$
- 1: Encode given $\{v_k\}$ and $\{\gamma_k\}$:
 $\alpha(u) = \operatorname{argmin} \{d(u, v_k) + \lambda|\gamma_k|\}$
- 2: Update VLC given $\alpha(u_k)$ and $\{v_k\}$
 $|\gamma_k| = -\log P(\alpha(u)=k)$
- 3: Update set of reconstruction levels given $\alpha(u_k)$ and $\{v_k\}$
 $v_k = \operatorname{argmin} E \{d(u, v_k) \mid \alpha(u)=k\}$
- 4: Repeat steps 1 - 3 until convergence

* 1989, has been proposed for Vector Quantization



Entropy-Constrained Scalar Quantization: High Resolution Approximations

- Assume: uniform quantization: $P_k = f_k \Delta$

$$\begin{aligned}
 R &= -\sum_{k=0}^{N-1} P_k \log P_k = -\sum_{k=0}^{N-1} f_k \Delta \log(f_k \Delta) \\
 \sum \Delta &\approx \int du \left| \begin{aligned} &= -\sum_{k=0}^{N-1} f_k \Delta \log(f_k) - \sum_{k=0}^{N-1} f_k \Delta \log(\Delta) \\ &\approx \underbrace{\int_{\mathbb{R}} f_U(u) \log(f_U(u)) du}_{\text{Differential Entropy } h(U)} - \log \Delta \underbrace{\int_{\mathbb{R}} f_U(u) du}_1 \\ &= h(U) - \log \Delta \end{aligned} \right.
 \end{aligned}$$

- Operational distortion rate function for Gaussian RV

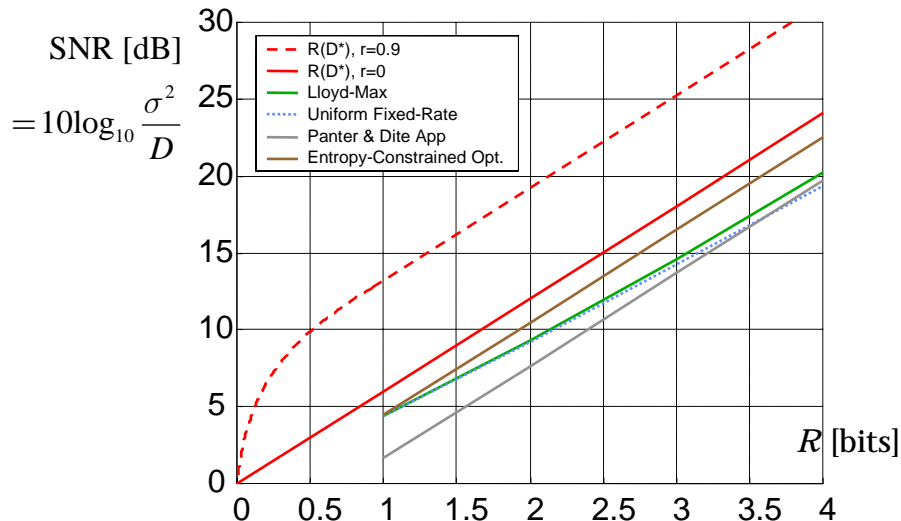
$$U \sim N(0, \sigma^2), \quad D(R) \approx \frac{\pi}{6} \frac{e}{\sigma^2} 2^{-2R}$$

- It can be shown that for high resolution:

Uniform Entropy-Constrained Scalar Quantization is optimum



Comparison for Gaussian Sources

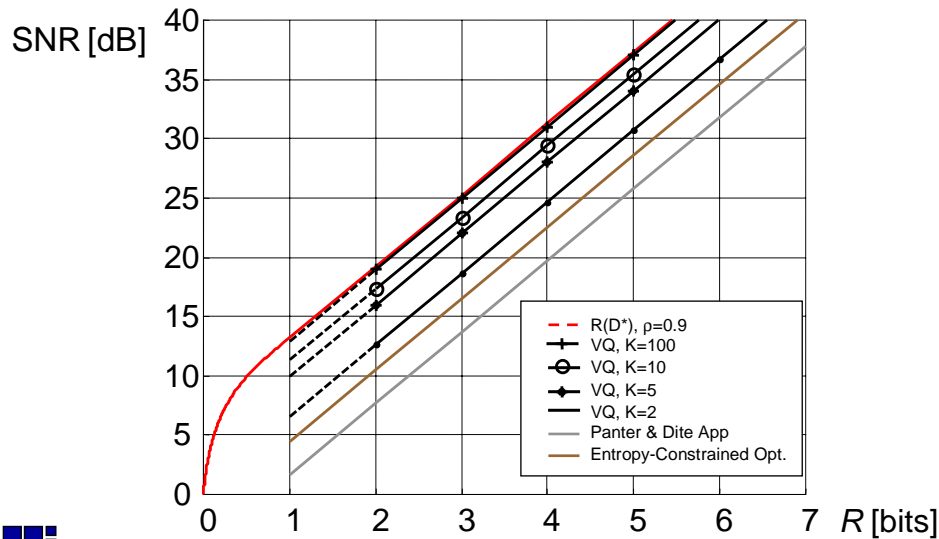


Vector Quantization

- So far: scalars have been quantized
- Encode vectors, ordered sets of scalars
- Gain over scalar quantization (Lookabaugh and Gray 1989)
 - Space filling advantage
 - Z lattice is not most efficient sphere packing in K -D ($K > 1$)
 - Independent from source distribution or statistical dependencies
 - Maximum gain for $K \rightarrow \infty$: 1.53 dB
 - Shape advantage
 - Exploit shape of source pdf
 - Can also be exploited using entropy-constrained scalar quantization
 - Memory advantage
 - Exploit statistical dependencies of the source
 - Can also be exploited using DPCM, Transform coding, block entropy coding

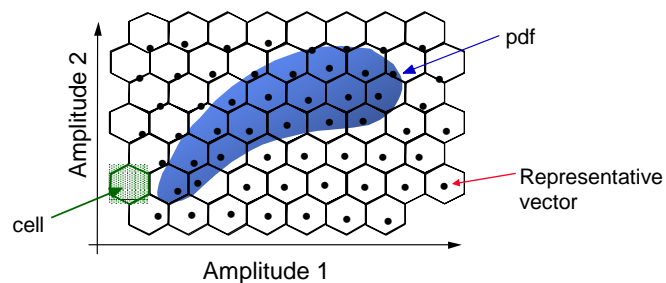


Comparison for Gauss-Markov Source: $\rho=0.9$



Vector Quantization II

- Vector quantizers can achieve $R(D^*)$ if $K \rightarrow \infty$
- Complexity requirements: storage and computation
- Delay
- Impose structural constraints that reduce complexity
- Tree-Structured, Transform, Multistage, etc.
- Lattice Codebook VQ



Summary

- Rate-distortion theory: minimum bit-rate for given distortion
- $R(D^*)$ for memoryless Gaussian source and MSE: 6 dB/bit
- $R(D^*)$ for Gaussian source with memory and MSE: encode spectral components independently, introduce white noise, suppress small spectral components
- Lloyd-Max quantizer: minimum MSE distortion for given number of representative levels
- Variable length coding: additional gains by entropy-constrained quantization
- Minimum mean squared error for given entropy: uniform quantizer (for fine quantization!)
- Vector quantizers can achieve $R(D^*)$ if $K \rightarrow \infty$ - Are we done ?
- No! Complexity of vector quantizers is the issue

Design a coding system with optimum rate distortion performance, such that the delay, complexity, and storage requirements are met.

